

Circuit QED with Carbon-Nanotube-Based Superconducting Quantum Circuits

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Hybrid circuit quantum electrodynamics (QED) involves the study of coherent quantum physics in solid state systems via their interactions with superconducting microwave circuits. Here we present **a crucial step in the implementation** of a hybrid superconducting qubit that employs a carbon nanotube as a Josephson junction. We realise the junction by contacting a carbon nanotube with a superconducting Pd/Al bi-layer, and implement voltage tunability of the **quantum circuit's frequency** using a local electrostatic gate. We demonstrate a strong dispersive coupling to a coplanar waveguide resonator by investigating the gate-tunable resonator frequency. We extract qubit parameters from spectroscopy using dispersive readout and find qubit relaxation and coherence times in the range of 10 – 200 ns.

Circuit quantum electrodynamics (cQED) with superconducting circuits [1] is a powerful platform used in on-chip quantum optics and quantum information processing [2]. Hybrid superconducting circuits provide access to coherent quantum properties of other systems based on their interactions with microwave photons or artificial atoms [3–5]. In recent years, a variety of hybrid superconducting qubits have been realised by replacing the conventional aluminium (SIS) Josephson junctions (JJ) with semiconductor-based (SNS) JJs, such as InAs nanowires [6, 7], InGaAs heterostructures [8] and graphene [9, 10]. For these SNS JJs the normal- or semiconductor is contacted with a superconducting material enabling a supercurrent to flow due to the superconducting proximity effect [11]. Cooper pair transport in such devices is described by Andreev reflections [12–14]. The conductance of semiconductors can be adjusted by applying a voltage to a nearby gate-electrode, which tunes the Cooper-pair transport and hence the Josephson energy of the junction.

A strong technical motivation for these new semiconductor-superconductor hybrid JJ qubits is to realise gate voltage tunable qubits and hence eliminate decoherence due to magnetic flux noise. Further, electric fields are much easier to localise compared to magnetic fields, which makes complex multi-qubit devices simpler to engineer. Additionally, qubit operation in moderate magnetic fields, for example to explore interactions with different spin systems, can be made possible due to the robustness of these hybrid JJs to magnetic field [9, 15].

An interesting material to use in a JJ is the carbon nanotube (CNT), which can display ballistic electronic transport, and clean signatures of Andreev reflections

when contacted with superconductors [16]. Using a CNT as the junction allows to make use of its exceptional mechanical properties, which could offer a potential platform for creating quantum interference between a qubit and mechanical motion [17]. Further, ultra-clean CNTs offer ballistic transport characteristics [18], which could provide JJs with lower defect density as opposed to conventional Al JJs with an amorphous tunnel barrier. This might have a potential positive impact on qubit coherence via elimination of two-level fluctuator defects in the amorphous tunnel barrier oxide [19–21]. Recent progress in CNT fabrication techniques might allow a CMOS-like design flow and processing of CNT JJs, which are free of nanoscale imperfections [22]. Hybrid devices incorporating proximitised CNTs allow the study of Andreev levels [16, 23–25] and they are also predicted to carry Majorana fermions [26–28], which could be beneficial for topological quantum computing [29].

In the work presented here proximitised CNTs are used as the JJ in a common planar 2D superconducting qubit architecture and their performance as a qubit is analysed via a coupled microwave resonator. Resonator and qubit spectroscopy are performed as a function of applied gate voltage and strong dispersive coupling on the order of 100 MHz is observed. Power-dependent qubit spectroscopy is used to extract the likely Josephson energy E_J and transmission T of the qubits' CNT junctions within a simple few-channel junction model [30]. Further, **tentative evidence of Rabi oscillations is observed. Hence,** the coherence is investigated using a pulse chopping technique [31, 32] and qubit spectroscopy [33, 34], allowing T_1 and T_2 times in the range of 10-200 ns to be observed.

Fig. 1 shows images and a circuit diagram of the device studied. Device fabrication begins with CNTs grown via chemical vapour deposition on a Si/SiO₂ (450 μm/300 nm) substrate. The microwave resonators are then patterned via electron beam lithography (EBL). Prior to metal deposition an oxygen plasma etch is carried out to remove any CNTs that might electrically short

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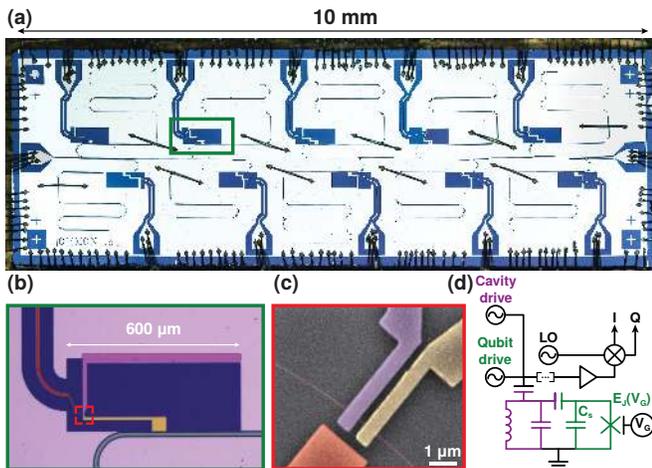


Figure 1. Carbon nanotube superconducting quantum circuit device. (a) Optical image of the device chip. A coplanar microwave transmission line in the centre addresses 10 multiplexed $\lambda/4$ resonators with different resonance frequencies. Each resonator has a cut-out in the ground plane close to its electric field anti-node for qubit fabrication and a single DC line allowing voltage tunability of the qubit frequency. (b) False colour optical image of a single qubit with the island (yellow) capacitively coupled to the resonator (green) and the other side shorted to ground (purple). The island capacitively shunts the CNT JJ to the surrounding ground plane. A side gate (red) is used to tune the circuit's frequency. (c) False colour SEM image of a CNT (pink) contacted with two superconducting contacts (yellow/purple) separated by 300 nm and a side gate (red). (d) Electrical circuit diagram of the device along with a sketch of its readout and control circuitry. The qubit (green) is capacitively coupled to the resonator (purple). A side gate with applied voltage V_G tunes the Josephson energy E_J of the qubit. The resonator itself is capacitively coupled to a transmission line which is used to send microwave tones to the qubit and its response is measured using a standard heterodyne detection scheme.

the microwave circuits. Afterwards, 100 nm of Al is deposited via electron beam evaporation. Following lift-off, SEM imaging is used to locate and select CNTs for the qubits. The contacts to the CNT and island of the qubit are then patterned with EBL, post-development cleaned using UV ozone and metalised with a Pd/Al (4/80 nm) bi-layer. Before sample mounting, the room-temperature resistances of the CNT JJs are measured to check the fabrication yield. Roughly 80% of the fabricated devices conduct at room-temperature and exhibit resistances between $7 \text{ k}\Omega < R_n < 100 \text{ k}\Omega$. A series of 3 chips were fabricated, each consisting of 10 potential qubits. Each chip contains 10 $\lambda/4$ resonators with different frequencies, multiplexed via capacitive coupling to a single microwave transmission line (Fig. 1 (a)). Close to each resonator's electric field anti-node, qubits are fabricated and a dedicated DC electrostatic gate is used for control of the chemical potential of the CNT (Fig. 1 (b-c)).

The system is cooled below 20 mK in a dilution refrigerator and measured using standard cQED measurement

techniques, see Fig. 1 (d). First, the transmission spectrum of a device is measured via the feedline to identify the individual resonance frequencies of the 10 resonators, at each of which a narrow ($\sim 1 \text{ MHz}$) absorption dip is observed. Subsequently, S_{21} spectroscopy as a function of gate voltage V_G of each individual resonance is performed and resonators which exhibit a clear gate-dependent resonance frequency are selected for further investigations as these potentially correspond to working CNT-qubits. Usually 20-50% of all devices on one chip show this dependence. From these, two devices showing similar gate-dependent behaviour are carefully characterised, hereafter labeled as device QA and QB.

Fig. 2 (a) shows resonator spectroscopy as a function of DC voltage applied to the gate electrode (V_G) on device QA. At a single gate-voltage V_G , an absorption line corresponding to the resonator is observed (inset Fig. 2 (b)). Tuning V_G the absorption line moves in frequency, exhibiting a broad gate region ($V_G < -30 \text{ V}$) where it is approximately constant, indicating the bare resonator frequency of $f_0 = 5.572 \text{ GHz}$, and another region ($V_G > -20 \text{ V}$) where the resonance is quasi-periodically shifting to frequencies up to $\sim 10 \text{ MHz}$ higher than f_0 . In cQED this is indicative of the resonator being dispersively coupled to a circuit with tunable transition frequency $f_Q < f_0$. From here onwards we will refer to the circuit as qubit and will investigate the circuit's level of qubit behaviour.

In a second measurement, we carry out spectroscopy to identify the potential qubit's frequency. Here the cavity drive is set to track the particular resonance frequency f_r at each V_G and simultaneously a spectroscopic probe tone is fed onto the input line (qubit drive). While the qubit drive is swept in frequency the amplitude response at f_r is measured. This measurement performed on device QA is presented in Fig. 2 (b). A spectroscopic response is observed for $V_G > -30 \text{ V}$. The values of V_G exhibiting a spectroscopic response coincide exactly with the values exhibiting shifts in f_r , cf. Fig. 2 (a). This is consistent with the dispersive regime of cQED [35] indicating the presence of a qubit, frequency tunable between $2.8 \text{ GHz} < f_Q < 4.2 \text{ GHz}$. The data in Fig. 2 (b) likely shows the $f_{01} = f_Q$, i.e. ground to first excited state transition of the qubit as a function of gate voltage. Similar measurements were also performed on device QB, see supplementary material [36] and Table I.

We begin our analysis by using a simple model of a two-level system dispersively coupled to a harmonic oscillator. In this model the coupling strength g between the resonator and the qubit can be estimated using the

	f_0 [GHz]	f_Q [GHz]	g_{\max} [MHz]	χ_{\max} [MHz]
Qubit A	5.572	2.8 – 4.2	113	10
Qubit B	4.595	2.4 – 3.5	85	7

Table I. Parameters extracted from resonator and qubit spectroscopy measurements for devices QA and QB.

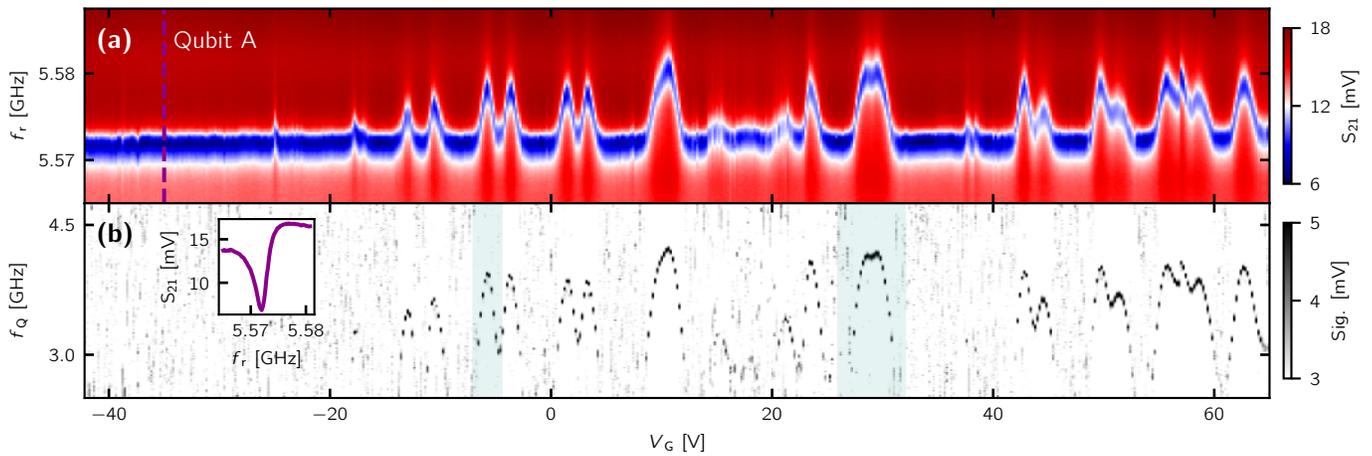


Figure 2. Resonator and qubit spectroscopy as function of applied gate voltage V_G for device QA. The transmission S_{21} is expressed in terms of the corresponding demodulated signal voltage. (a) Resonator spectroscopy of device QA as a function of V_G . (b) Qubit spectroscopy of device QA as a function of V_G . The shaded areas mark the features used for qubit analysis (cf. Fig. 3, V_G is shifted due to drift). Inset: Line-cut through spectrum indicated in (a).

dispersive shift χ of the resonator [33, 37]. To first order, due to the coupling to the f_{01} transition, $\chi = \frac{g^2}{\Delta}$, where $\Delta = f_r - f_Q$ is the detuning between the qubit and the resonator [37]. Using this expression and the data of QA (Fig. 2 (a) and (b)) yields an estimate of the coupling strength in the range $48 < g < 113$ MHz that increases with observed f_Q , see supplementary material [36]. This behaviour is expected for transmon qubits because $g \propto E_J^{1/4}$ and $f_Q \propto E_J^{1/2}$ [37]. Similar values are observed with QB, summarised in Table I.

In reality our qubits are not expected to be pure two-level systems but rather anharmonic oscillators with many levels [37], and hence exhibit more than one transition. Therefore, mapping out f_{01} is the first step of characterising the voltage tunable qubit. To fully describe the qubit, the Josephson energy E_J , charging energy E_C , transmission T and the number of conduction channels N are needed. Electrostatic finite element simulations for the exact qubit designs yield values for the shunt capacitance to ground and the island's capacitance to the resonator (cf. Fig. 1 (b)), both contributing to the overall capacitance C_Σ necessary to calculate $E_C = e^2/2C_\Sigma$ ($E_C^{QA} = 508$ MHz, $E_C^{QB} = 391$ MHz). Nonetheless, the other parameters related to the JJ cannot be deduced from the qubit's fundamental frequency alone. Here, we show that they can be extracted by measuring the qubit at different drive powers. For this purpose, we use the same qubit spectroscopy technique as in Fig. 2 (b), holding V_G fixed and varying the qubit drive power. Such a measurement is presented in Fig. 3 (a). At a low drive power, $P = -45$ dBm at the output of the microwave generator, only a single peak is observed in the qubit spectroscopy (Fig. 3 (a) bottom trace). If the power of the qubit drive is increased to $P = -30$ dBm, a more complicated multi-peak response is observed, which exhibits a second peak at frequencies just below f_{01} , see

Fig. 3 (a). This is indicative of a weakly anharmonic circuit, such as a transmon qubit, where the lower peak corresponds to the two photon transition from the ground state to the second excited state, i.e. $f_{02}/2$. Note that this second spectral peak is not always clearly present in our data on these devices, and a broader spectral feature is consistently seen at higher drive powers, which it is not possible to resolve into clear individual peaks. This may be due to significant charge dispersion of higher qubit energy levels, or other sources of decoherence. We measure the frequency of the two clearest spectral lines over a range of V_G (indicated in Fig. 2 (b)) from data similar to that seen in Fig. 3 (a). Interpreting them as the f_{01} and $f_{02}/2$ transitions of a qubit with at least three energy levels, we can extract a possible anharmonicity between these three levels as $\alpha = 2(f_{01} - f_{02}/2)$.

CNT JJs and other types of JJs, such as weak links made from narrow superconducting constrictions, normal metal or a semiconductor, have energy phase relations that differ from standard SIS JJs [38–41]. Assuming the CNT JJs with a channel length of 300 nm are in the short junction regime, cooper pair transport is mediated by Andreev bound states [41–43]. Hence, we cannot interpret the measurements by using a standard Cooper pair box (CPB) Hamiltonian, but rather by using a second order perturbation theory approach to a modified CPB Hamiltonian presented by *Kringhøj et al.* [30]. Here, α and f_{01} are given by

$$h\alpha = E_C \left(1 - \frac{3T}{4} \right) - \frac{E_C^{3/2}}{\sqrt{2E_J}} \left(1 - \frac{15}{4} T \left(1 - \frac{3T}{4} \right) \right), \quad (1)$$

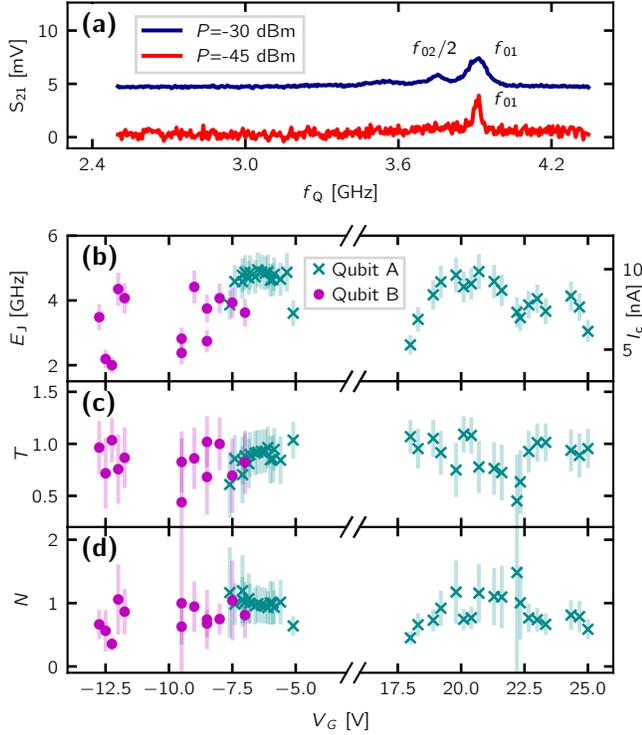


Figure 3. Qubit spectroscopy, qubit parameter E_J and junction characteristics T and N . (a) Qubit spectroscopy traces at two different qubit drive powers for device QA, offset for clarity. (b) Extracted values for E_J as a function of V_G for QA (cyan crosses) and QB (purple points). The critical current I_c , calculated from E_J is indicated on the right y-axis. (c) Extracted values for T as a function of V_G for QA and QB. (d) Values of N calculated from T .

$$hf_{01} = \sqrt{8E_J E_C} - E_C \left(1 - \frac{3T}{4}\right) + \frac{E_C^{3/2}}{2\sqrt{2E_J}} \left(1 - \frac{15}{4}T \left(1 - \frac{3T}{4}\right)\right), \quad (2)$$

where it is assumed that all conduction channels exhibit equal transmission T , see supplementary material for details [36]. Using Eq. 1 and Eq. 2 together with the values extracted for f_{01} , α and E_C , the parameters E_J and T can be calculated by solving the resulting set of equations, yielding a deeper understanding of the qubit's CNT JJ.

In Fig. 3 (b) the Josephson energy E_J , extracted with the method mentioned above, is presented as a function of gate voltage V_G (note a small gate drift compared to Fig. 2 (b)). The error bars are determined by the measurement error of f_{01} and $f_{02}/2$, as the qubit power spectroscopy data can be noisy and sometimes exhibit a complicated multi-peak structure, making the peak distinction difficult. The Josephson energy can also be used to calculate the critical current I_c of the qubit's JJ using $E_J = \hbar I_c / (2e)$, see Fig. 3 (b). We find $4 \text{ nA} < I_c < 10 \text{ nA}$, with an average value $\langle I_c \rangle = 8.2 \pm 1.6 \text{ nA}$ across both qubit devices. These values of I_c are comparable to $1 \text{ nA} < I_c < 17 \text{ nA}$ that we independently observed in

DC bias spectroscopy of CNT JJs contacted with identical processing. The extracted values of T are presented in Fig. 3 (c). We consistently find $0.33 < T < 1$ with a mean value of $\langle T \rangle = 0.85 \pm 0.16$, indicating high quality contacts to the CNT. Further, the extracted values of T can be used to calculate the number of channels N contributing to transport via $E_J = \Delta NT/4$ [30], where Δ is the induced superconducting gap of the CNT JJ (we estimate $\Delta = 90 \pm 10 \mu\text{eV}$ using DC bias spectroscopy measurements on CNT devices contacted with **identical processing** [44]), see Fig. 3 (d). The calculated values for N congregate around $N = 1$, indicating that only one conduction channel is strongly coupled to the superconducting contacts. This is in strong contrast to conventional aluminium SIS JJs, where $N \gg 1$ and $T \ll 1$.

The obtained values of E_J in conjunction with the simulated values of E_C can be used to calculate the ratio E_J/E_C . We find a mean $E_J/E_C \approx 9 \pm 2$ for both QA and QB. These values agree well with electrostatic simulations of the device design, yielding $E_J/E_C \sim 12$, giving confidence in the extraction method of E_J and T . Note that the estimated average E_J/E_C ratio places the qubit between the transmon regime ($E_J/E_C > 20$) and the Cooper pair box regime ($E_J/E_C < 1$), where a low E_J/E_C leads to a complex energy level structure with charge dispersion making the qubit susceptible to charge noise [37]. However, charge dispersion is predicted [45] and shown [46] to vanish in channels where the transmission is approaching unity, which is the case for JJs made from CNTs.

We finally report on investigations of the relaxation and coherence times of the CNT-qubit devices. Time domain measurements with a weak continuous readout drive [47, 48] showed evidence of Rabi oscillations [36]. In Fig. 4 (a) the extracted Rabi oscillation frequency Ω is shown as a function of the qubit drive amplitude. For drive amplitudes $< 0.5 \text{ V}$, Ω increases linearly with a slope of $\sim 78.3 \text{ MHz/V}$, which is a characteristic feature of Rabi oscillations. For pulse amplitudes $> 0.5 \text{ V}$, Ω saturates at around $\sim 38 \text{ MHz}$, which may be explained as being due to the low anharmonicity of the qubit, i.e. the qubit will be driven into higher energy states at drive rates greater or equal to the anharmonicity. It is worth mentioning that the decay of the observed oscillations yielded a time constant of $\sim 50 \text{ ns}$, which is consistent with the coherence measurements presented below. **The experimental setup was not suitable to explicitly resolve such fast decaying Rabi oscillations. Hence, for early qubit analysis** we employ an alternative pulsed technique for measuring the relaxation time, previously used in quantum dot charge qubits [31, 32]. The method was tested on conventional superconducting qubits to confirm that it yields the same result as standard techniques (see supplementary material [36]).

To measure T_1 , a pulse chopping method [31, 32] is used. The resonator is continuously measured with a weak cavity drive at f_0 for a time of $100 \mu\text{s}$. Simultaneously, the qubit is driven on resonance with a pulse

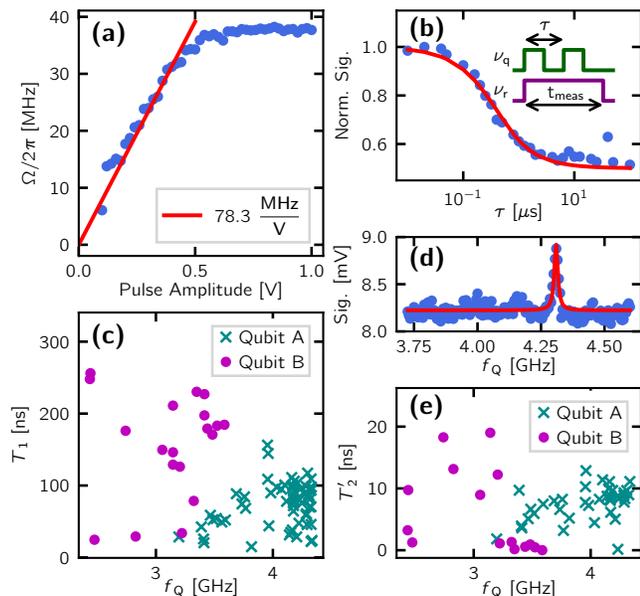


Figure 4. Signatures of Rabi oscillations and T_1 and T_2' measurements. (a) Rabi oscillation frequency $\Omega/2\pi$ as a function of the qubit pulse amplitude. For a pulse amplitudes < 0.5 V, $\Omega/2\pi$ is fit with a linear function (solid red line) yielding a slope of 78.3 MHz/V. (b) Data of a single T_1 experiment on device QA. The measurement response is fitted to Eq. 3 (red curve), yielding $T_1 = 117.3 \pm 5.8$ ns. Inset: Pulse scheme (qubit pulse - green, measurement pulse - purple). (c) Measured T_1 as a function of f_Q for devices QA (crosses) and QB (circles). (d) A single measurement of T_2' on device QA. Qubit spectroscopy trace is fitted to Eq. 4 (red curve), yielding $T_2' = 19.0 \pm 1.5$ ns. (e) Measured T_2' as a function of f_Q for devices QA (crosses) and QB (circles).

train of 50% duty cycle, and for each measurement the pulse period τ is varied, see inset Fig. 4 (b). For very short τ , i.e. $\tau \ll T_1$, the qubit drive randomises the qubit between the ground and first excited state and it has no time to relax. In the case of very long τ , i.e. $\tau \gg T_1$, the qubit has time to relax to the ground state in between drive pulses. Therefore, in the latter case, the measured signal is the time average of the qubit being in the ground and excited state, giving a signal of half the value found in the limit $\tau \rightarrow 0$. A measurement following this procedure is presented in Fig. 4 (b). The data is normalised with respect to a measurement with the qubit drive turned off and fitted to

$$S(\tau) = \frac{1}{2} + \frac{T_1(1 - e^{-\tau/(2T_1)})}{\tau} \quad (3)$$

where T_1 is the only free parameter. In the measurement shown in Fig. 4 (b) the fit yields $T_1 = 117.3 \pm 5.8$ ns. The relaxation time T_1 was measured across a range of gate voltages, and therefore a range of f_Q , for both devices, see Fig. 4 (c). QB on average exhibits a longer $\langle T_{1,QB} \rangle = 151 \pm 71$ ns than QA where $\langle T_{1,QA} \rangle = 74 \pm 30$ ns. The longest T_1 values were 250 ns and 150 ns for QB and QA respectively.

A lower bound, T_2' , on qubit coherence time T_2 can be found by measuring the linewidth of a low power qubit spectroscopy trace [33, 34], see Fig. 4 (d). Fitting a Lorentzian with linewidth $2\delta\nu_{\text{HWHM}}$ to the qubit transition peak, T_2' can be calculated via

$$2\pi\delta\nu_{\text{HWHM}} = \frac{1}{T_2'} = \left(\frac{1}{T_2'^2} + n_s\omega_{\text{vac}}^2 \frac{T_1}{T_2'} \right)^{1/2} \quad (4)$$

where $n_s\omega_{\text{vac}}^2$ is proportional to the microwave input power, ω_{vac}^2 the vacuum Rabi frequency, and n_s the number of photons in the resonator [33, 34]. Hence, at low qubit drive powers the linewidth should be the least broadened. Here, low power corresponds to the lowest qubit drive power which still resulted in a visible qubit spectroscopy peak. We assume this power corresponds to $n_s \approx 0$. However, it is important to note that we specifically quote T_2' as a lower limit for T_2 and hence n_s does not need to be known. This results in $T_2' = 19.0 \pm 1.5$ ns for the data presented in Fig. 4 (d).

The measurement and its analysis was repeated for different qubit frequencies f_Q , see Fig. 4 (e). While T_2' seems to increase slightly with increasing f_Q for QA, this is not true for QB. Coherence is highest at around $f_Q = 3$ GHz, with $T_2' = 25$ ns for QB, but significantly reduced at $f_Q = 3.5$ GHz. On average QA exhibits a longer $\langle T_{2',QA} \rangle = 10 \pm 5$ ns compared to QB $\langle T_{2',QB} \rangle = 6 \pm 6$ ns. We stress that these coherence times only represent a lower bound for T_2 .

Decoherence and fast relaxation in these devices could be attributed to dissipation due to dirty, disordered CNTs, Purcell decay into the gate line, strong dielectric loss due to the thick SiO_2 , and residual resistance to the superconducting leads. The loss tangent of thermal SiO_2 was measured to be $\tan \delta \sim 3e^{-4}$ [49], limiting the qubit's relaxation time to $T_1 \sim 1 \mu\text{s}$ within the accessible frequency range. Dissipation in nanoscale weak link JJ oscillators, made from aluminium, was previously mentioned as a possible source of decoherence [50–52]. Additionally, the short T_2' could also stem from Andreev levels in the junction interacting with acoustic phonons [53–55].

The experiments described here **demonstrate crucial steps** in the implementation of a voltage tunable superconducting qubit based on a CNT JJ. The device is of similar geometry and exhibits similar gate voltage behaviour to previously reported voltage tunable superconducting qubit devices [6–10]. Simultaneous resonator and qubit spectroscopy showed clear evidence of qubit-resonator coupling with coupling strength on the order of $g \sim 100$ MHz, comparable to cQED experiments with conventional transmon qubits. Qubit spectroscopy at high drive powers was used to extract the qubit parameter E_J and ratio E_J/E_C as well as I_c and T of the CNT JJ. From the values for T , values of $N \sim 1$ were calculated, indicating that only one conduction channel is strongly coupled to the superconducting leads of the qubit. Further, **evidence of Rabi oscillations as well as** qubit relaxation and coherence times in the range

10–200 ns were observed. **It remains to explicitly demonstrate Rabi oscillations and full qubit control.**

Advances in fabrication, e.g. using suspended ultra-clean CNT JJs could lead to significant improvements to the coherence times of these devices. Such JJs have already been individually realised [56]. Fabrication improvements in other hybrid qubit designs, such as those based on InAs nanowires resulted in $T_1 = 5 - 20 \mu\text{s}$ [15, 57], similar to state-of-the-art, aluminium-based flux-tunable transmon qubits [58]. The implementation of a superconducting **quantum circuit based on** a CNT presented here offers potential for unique experiments in order to create quantum interference between a qubit and mechanical motion [17]. Additionally, CNT-based qubits could be used as ultra-sensitive force sensors [59] and if arranged in a SQUID geometry as a de-

tor for magnetic moments [38]. Furthermore, these qubits based on proximitised CNTs could be utilised to study Andreev physics [23–25] and investigate the prediction of carrying Majorana fermions [26–28], which could be valuable for topological quantum computing.

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